

Problem 1 (2.5 points)

In a nonmagnetic, lossy dielectric medium, a 400 MHz plane wave is characterized by the magnetic field phasor:

$$\tilde{\mathbf{H}} = (\hat{\mathbf{x}} - j4\hat{\mathbf{z}}) e^{-0.05y} e^{-j10y}$$

Obtain time-domain expressions for: the electric and magnetic field vectors.

- the electric field vector (1 points)
- the magnetic field vector (1 points)
- Determine the polarization of the wave (0.5 point)

(Free space permittivity: $\epsilon_0=8.85 \times 10^{-12}$ F/m, free space permeability: $\mu_0=4\pi \times 10^{-7}$ H/m)

Solution:

From the expression of the magnetic field phasor, the attenuation and phase constants are known:

$$\alpha = 0.05 \text{ Np/m}$$

$$\beta = 10 \text{ rad/m}$$

Frequency $f=500 \text{ MHz}=5 \times 10^8 \text{ Hz}$ and the propagation direction is along y axis. So:

$$-\omega^2 \mu \epsilon' = -\omega^2 \mu_0 \epsilon_0 \epsilon_r' = \alpha^2 - \beta^2 \approx -100$$

Thus, the real part of the relative permittivity is:

$$\epsilon_r' = \frac{100}{\omega^2 \mu_0 \epsilon_0} = \frac{100}{(6.28 \times 4 \times 10^8)^2 \times 4\pi \times 10^{-7} \times 8.85 \times 10^{-12}} = 1.426$$

And for the imaginary part of the relative permittivity:

$$2\alpha\beta = \omega^2 \mu \epsilon'' = \omega^2 \mu_0 \epsilon_0 \epsilon_r''$$

So:

$$\epsilon_r'' = \frac{2\alpha\beta}{\omega^2 \mu_0 \epsilon_0} = \frac{2 \times 0.05 \times 10}{(6.28 \times 4 \times 10^8)^2 \times 4\pi \times 10^{-7} \times 8.85 \times 10^{-12}} = 0.014$$

Since $\epsilon_r'' / \epsilon_r' = 0.0098 \ll 1$, the medium has very low loss so we can use low loss approximation.

The intrinsic impedance is:

$$\eta = \sqrt{\frac{\mu}{\epsilon'}} = \sqrt{\frac{4\pi \times 10^{-7}}{1.426 \times 8.85 \times 10^{-12}}} = 315.5 \text{ } \Omega$$

Electrical field phasor is thus

$$\begin{aligned} \tilde{\mathbf{E}} &= -\eta \hat{\mathbf{k}} \times \tilde{\mathbf{H}} = -315.5 \hat{\mathbf{y}} \times (\hat{\mathbf{x}} - j4\hat{\mathbf{z}}) e^{-0.05y} e^{-j10y} \\ &= (j4\hat{\mathbf{x}} + \hat{\mathbf{z}}) 315.5 e^{-0.05y} e^{-j10y} \end{aligned}$$

In time domain:

$$\begin{aligned}
\mathbf{E}(t) &= \operatorname{Re}\{\tilde{\mathbf{E}}e^{j\omega t}\} \\
&= \hat{\mathbf{x}}1262e^{-0.05y} \cos(\omega t - 10y + 90^\circ) + \hat{\mathbf{z}}315.5e^{-0.05y} \cos(\omega t - 10y) \\
&= -\hat{\mathbf{x}}1262e^{-0.05y} \sin(\omega t - 10y) + \hat{\mathbf{z}}315.5e^{-0.05y} \cos(\omega t - 10y) \\
\mathbf{H}(t) &= \operatorname{Re}\{\tilde{\mathbf{H}}e^{j\omega t}\} \\
&= \hat{\mathbf{x}}e^{-0.05y} \cos(\omega t - 10y + 90^\circ) + \hat{\mathbf{z}}4e^{-0.05y} \sin(\omega t - 10y)
\end{aligned}$$

The polarization of the wave is right-hand elliptical.

Problem 2 (1.5 points)

A lossless 100Ω transmission line of $3\lambda/8$ in length is terminated in an unknown load impedance. The input impedance Z_{in} is measured to be $60 + j80$, determine the load impedance.

Solution:

Input impedance is:

$$Z_{in} = Z_0 \frac{z_L + j \tan \beta l}{1 + j z_L \tan \beta l} = 100 \frac{z_L + j \tan(3\pi/4)}{1 + j z_L \tan(3\pi/4)} = 100 \frac{z_L - j}{1 - j z_L} = 60 + j80$$

Solve for z_L and find $z_L = 3$, so the load impedance is 300Ω .